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# Design and Characterisation of VO<sub>2</sub> Based Switches for Ultra-Fast Reconfigurable Devices

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Abstract — Mott transition materials seem particularly promising to design reconfigurable devices. In this paper, Vanadium Dioxide (VO<sub>2</sub>) is studied and characterized to design switches and elementary RF devices. The study shows experimentally promising RF performances under electrical control and also for an optical excitation of the VO<sub>2</sub> with ultrashort switching time. These results put forward the use of VO<sub>2</sub> switches as a core function to design ultra-fast purpose RF devices dedicated to future military and civilian standards.

Keywords — Vanadium dioxide, ultra-fast switches, reconfigurable architectures, frequency shift keying.

#### I. INTRODUCTION

The new military and civilian standards (e.g. massive MIMO and 5G Network purpose beamforming) and the allocation of the new frequency bands had increased the needs of reconfigurable and agile radiofrequency (RF) devices. In these circumstances, the switching speed of the RF devices is an essential and crucial parameter. Actual devices are, for the most part, based on mechanical technologies like the Micro-Electro-Mechanical Switches (MEMS) or are based on semiconductors components like PIN Diodes, varactors or Field Effect Transistors (FET) [1]. The latter are compact and exhibit a switching time in the range of 50 ns down to 1 ns, but are less interesting over wide band applications where non-linear parasitic effects are hard to take into account. On the other hand, MEMS switching time ranges typically between 500 µs down to 1 µs [1]. In that context, the Vanadium Dioxide (VO<sub>2</sub>), which is a phase switching ultrawide band material [2], has been identified as a promising solution with a switching time in the range of 0.1 µs down to 100 ps [1].

This study focuses on the design of elementary agile RF devices using  $VO_2$  based switches. The RF performances are experimentally characterized under an electrical control and the very low switching time using an optical excitation. The design and complete characterization of these devices are essential to have a set of reliable elementary blocks in order to build more advanced functions (SP2T, SP4T, switched-line phase shifters ...).

#### II. VANADIUM DIOXIDE VO2: PROPERTIES AND ACTIVATIONS

The Vanadium Dioxide  $(VO_2)$  is a Metal-Insulator Transition (MIT) material also known as the Mott

Transition [3]-[4]. When increasing the temperature from ambient temperature, the  $VO_2$  is in its semiconductor state (insulator) until the insulator  $\rightarrow$  metal transition temperature ( $T_{cm} = 68^{\circ}\text{C}$ ), and becomes conductor above. Decreasing temperature, this reversible transition (metal  $\rightarrow$  insulator) occurs at a slightly lower temperature ( $T_{ci} = 65^{\circ}\text{C} < T_{cm}$ ) due to a well-known hysteresis properties.  $T_{cm}$ ,  $T_{ci}$  and electrical resistivity can be adjusted according to material properties.

The  $VO_2$  transition can be obtained either with a thermal [5], an optical [6] or an electrical [7] command.

The thermal command has the disadvantage of being too slow due to the classical thermal inertia needed to heat up and especially to cool down the VO<sub>2</sub>. Conversely, the electrical and optical commands are much faster (potentially under 1 ps for the optical command) [8] and therefore particularly relevant for the foreseen RF applications.

#### III. DESIGN AND IMPLEMENTATION

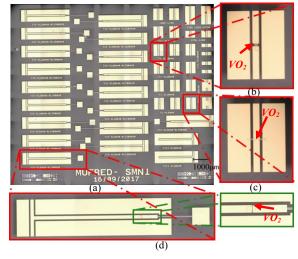
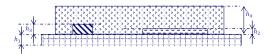


Fig. 1. (a) Photo of the wafer with the devices, (b) series switch, (c) shunt switch and (d) frequency tunable stub based on  $VO_2$ .

Several devices are studied in this article (Fig. 1(a)). Firstly, Coplanar Waveguide (CPW) switches are designed in series (Fig. 1(b), *i.e.* VO<sub>2</sub> thin film between input and output RF ports) and shunt (Fig. 1(c) with VO<sub>2</sub> between ground and hot lines of CPW) configurations.



Layer		-		
Material	R-cut Sapphire	Molybdene	$VO_2$	Gold Au
Thickness	$h_1 = 508 \mu m$	$h_2 = 150 \text{ nm}$	$h_3 = 200 \text{ nm}$	$h_4 = 1.5 \mu m$
Process		Lift-off	PLD	Lift-off
Lab		XLIM	IRCER	XLIM

Fig. 2. Cross-section (not to scale) of the different layer implementation.

Those devices are implemented with different sizes ( $VO_2$  gap length between 5  $\mu s$  to 30  $\mu s$ ) to enhance the understanding and the characterization of the material properties and the impact on RF performances.

In a complementary approach,  $VO_2$  material has also been integrated into a set of simple RF devices to make them agile. Several stub configurations have been designed: an open circuit one is detailed below (Fig. 1(d)). It consists of a  $VO_2$  based switch in series configuration in order to increase the length of the stub and consequently shift the resonance frequency toward the low frequency, *i.e.* a shift from 15 GHz to 13 GHz.

In that case, a Molybdene biasing line has been added and optimized to control VO<sub>2</sub> state through an electrical command without degrading the RF performances.

The main process of devices implementations are illustrated in Fig. 2 together with the material characteristics. Moreover, four probes measurements showed a  $R_{\rm OFF}/R_{\rm ON}$  ratio of  $2.5\times10^4$  after  $VO_2$  deposition on Sapphire using PLD.

#### IV. RF RESULTS

The first measurements have been conducted at Lab-STICC using a 200  $\mu m$  GSG pitched probe station and under voltage command applied between both ends of the  $VO_2$  material. Only, the results of most relevant devices (series switches and frequency tunable stubs) for our final applications are shown here for brevity.

#### A. Series Switches Specifications

The evolutions of the  $S_{11}$  and  $S_{21}$  parameters of the 5  $\mu m$  VO $_2$  gap based series switch are given in Fig. 3(a). In their insulator states (OFF state), all the VO $_2$  based switches understudied (gap of 5  $\mu m$ , 10  $\mu m$ , 20  $\mu m$  and 30  $\mu m$ ) show good isolation level  $S_{21} \leq$  -20 dB in the 8-18 GHz frequency band. In its metallic state (ON state), the VO $_2$  introduces variable losses with the gap length, i.e. VO $_2$  length. Those losses are consistent with VO $_2$  switch state-of-the-art, i.e. within -1.2 dB to -2.0 dB respectively for a gap varying from 5  $\mu m$  to 30  $\mu m$ , and with  $S_{11} \leq$  -14 dB over 8-20 GHz frequency band.

Like the resistivity versus temperature curve, I-V curve displays a hysteresis (Fig. 3(b)) with two different commutation voltages: insulator,  $V_{ci}$ , and metallic,  $V_{cm}$ . The voltage pair ( $V_{ci}$ ,  $V_{cm}$ ) is directly proportional to the length of the  $VO_2$  gap: (5.2 V, 9.6 V) and (7.9 V, 20.2 V) respectively for a 5  $\mu$ m and a 30  $\mu$ m gap. Hence a higher voltage is needed to obtain the same current while increasing the  $VO_2$  length (*i.e.* the corresponding resistance).

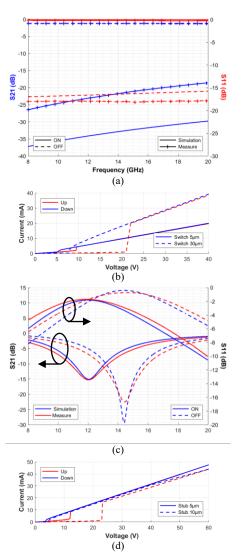


Fig. 3. (a) S-parameters of the 5  $\mu m$  VO $_2$  gap series switch, (b) I-V characteristic curve of the 5  $\mu m$  and 30  $\mu m$  VO $_2$  gap series switch (in series with a 2 k $\Omega$  resistor), (c) S-parameters of the 5  $\mu m$  VO $_2$  gap stub, (d) I-V characteristic curve of the 5  $\mu m$  and 10  $\mu m$  VO $_2$  gap stub (with a 1.3 k $\Omega$  resistive line).

At this stage on this wafer, 5 µm series switches seem to provide the best trade-off between isolation in OFF state, transmission losses in ON state and required switching power.

#### B. Stubs Specifications

To control the commutation of the  $VO_2$  when integrated into the frequency tunable stub, an additional DC probe has been used and placed over the biasing pad (Fig. 1(d)).

 $S_{11}$  and  $S_{21}$  parameters curves versus frequency are given are given for both states in Fig. 3(c) (VO<sub>2</sub> gap of 5 µm). A slight shift in the resonance frequency has been noticed toward the low frequency (14.4 GHz instead of 15 GHz) in the insulator state of the VO<sub>2</sub>, primarily explained by the R-cut Sapphire anisotropic permittivity. The frequency shift occurred by switching the VO<sub>2</sub> in its metallic state is also slightly higher than expected (2.4 GHz instead of 2 GHz) and can be mainly explained by the lower value of the resistive line's resistivity than used in the simulation. In both state,

good isolation parameters  $S_{21} \le -15 \text{ dB}$  are obtained at both work frequencies

A retro-simulation has allowed to obtain the effective permittivity of the R-cut Sapphire ( $\epsilon_r$  = 10.6) and Molybdene's resistivity (5.7×10<sup>-6</sup>  $\Omega$ .m).

The measured characteristic curves I-V are given in Fig. 3(d) for stubs with 5  $\mu m$  and 10  $\mu m$  VO $_2$  gap. Voltage pairs (3.7 V, 12.3 V) and (4.8 V, 23.3 V) respectively for 5  $\mu m$  and 10  $\mu m$  gaps are noticed. Once more, 5  $\mu m$  VO $_2$  gap provides the more relevant trade-off.

#### V. SWITCHING TIME AND OPTICAL COMMAND

Experimentations based on an optical command have also been conducted on 10  $\mu m$  series switches at Thales Research & Technology (TRT) on the same wafer by using a three wavelength laser (UV 355 nm, G 532 nm and NIR 1064 nm) with a 4 ns long pulse placed over a probe station with three optical lenses (×2, ×10 and ×50).

First, increasing the power of the excitation leads to a longer time in the metallic state (Fig. 4(a)) corresponding to the time the material needs to dissipate the additional energy provided by the laser. A deeper insight into the voltage at the  $VO_2$ 's terminals shows a step during the rising slope (Fig. 4(b)). The first slope corresponds to a fast reaction of the material but with an insufficient level of energy to maintain a long-lived conductivity. And the second one corresponds to a slower reaction where the energy is sufficient enough to reach conductivity as explained in [9].

Boltzmann sigmoid equation (1) has been used to estimate the rising time, where a and b are respectively the low and high horizontal asymptotes,  $\tau$  the time constant and  $t_0$  the central value.

$$f(t) = b + \frac{a - b}{1 + e^{\frac{t - t_0}{\tau}}} \tag{1}$$

(Model<sub>1</sub>) considers only the fast reaction and estimates a 4.4 ns rising time whereas (Model<sub>2</sub>) approximates both reactions and estimates a 33 ns rising time. Finally, a more complete model (Model<sub>3</sub>), taking into account the response disruption in the reaction provides a 40.1 ns rising time. This approach is of prime interest to identify the different phenomena that rule of the switching time and to quantify the latter in a real RF switch implementation. From these results, complementary studies are being conducted in order to decrease the switching time.

Moreover, the  $VO_2$  based series switches are observed under a microscope: before any optical excitation, the  $VO_2$  appears blue (Fig. 4(c)). After several excitations using the NIR wavelength, degradations are visible (Fig. 4(d)) and electrical parameters change irreversibly (Fig. 4(a)). This phenomenon might be explained by a partial ablation of the material by the laser excitation.

#### VI. CONCLUSION

Promising results are shown about RF specifications and rising time of the VO<sub>2</sub> to be used in the design of RF devices for ultra-fast reconfigurable applications.

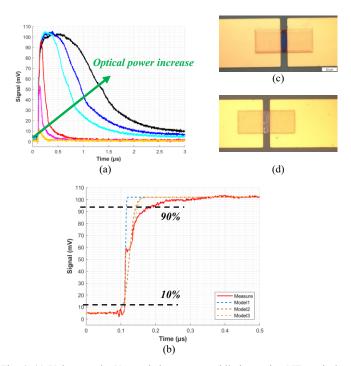


Fig. 4. (a) Voltage at the  $VO_2$  switches outputs while increasing NIR optical power and (b) model estimation of the switching time (10-90% rising time) with NIR wavelength, (c) Picture of the 10  $\mu$ m gap  $VO_2$  series switch before NIR optical excitation, (d) and after several NIR optical excitations.

Thanks to these experimental validations,  $VO_2$  switches implementation and optical control are currently under improvement to design ultra-fast SP2T and SP4T optically controlled microwave switches dedicated to advanced functions such as 2 and 3-bits phase shifters.

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